

Fig. 3. The  $1/\sqrt{\epsilon_{eff,f}}$  for the microstrip line on sapphire substrate.  $\epsilon_x^* = 9.4$ ,  $\epsilon_y^* = 11.6$ ,  $\gamma = 0$  in Fig. 1 — formula  $K(4)$ , — formula  $Y(8)$ , — formula  $YK(10)$ , — theoretical results; [4] for  $w/h = 0.1, 1, 10$ , and [5] for the other two dotted lines "x" experimental results [3],  $\bullet$   $f_i(11)$ ,  $\blacktriangle$   $f_K(5)$

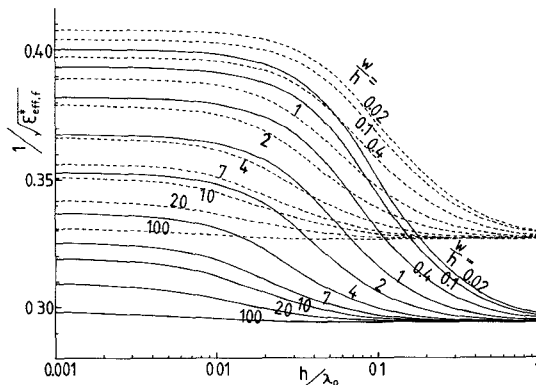


Fig. 4. The  $1/\sqrt{\epsilon_{eff,f}}$  for the microstrip line on sapphire substrate calculated by formula  $YK(10)$ . —  $\epsilon_x^* = 9.4$ ,  $\epsilon_y^* = 11.6$ ; - - -  $\epsilon_x^* = 11.6$ ,  $\epsilon_y^* = 9.4$ .

shows good agreement with the theoretical and experimental results and provides good design data. Fig. 4 shows the numerical results of formula  $YK$ . The equations also show the influence of the cutting angle of substrate on the dispersion properties.

#### IV. CONCLUSION

Three simple approximate dispersion formulas ( $K$ ,  $Y$ ,  $YK$ ) have been derived. The results obtained have been compared with other available results with good agreement. We have found that formula  $YK$  gives good design data.

The author thanks Prof. Naito, Tokyo Institute of Technology, for giving the copy of [8].

#### REFERENCES

- [1] P. H. Ladbrooke, "Some effects of field perturbation upon cavity-resonance and dispersion measurements on MIC dielectrics," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-25, pp. 892–903, Nov. 1977.
- [2] M. Horno, "Quasistatic characteristics of microstrip on arbitrary anisotropic substrates," *Proc. IEEE*, vol. 68, pp. 1033–1034, Aug. 1980.
- [3] T. C. Edwards and R. P. Owens, "2–18-GHz dispersion measurements on 10–100- $\Omega$  microstrip lines on sapphire," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-24, pp. 506–513, Aug. 1976.
- [4] Y. Hayashi and T. Kitazawa, "Analysis of microstrip transmission line on a sapphire substrate," *Trans. Inst. Electron. Comm. Eng. Japan*, vol. 62-B, pp. 596–602, June 1979, (in Japanese).
- [5] A.-M. A. El-Sherbiny, "Hybrid mode analysis of microstrip lines on anisotropic substrates," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-29, pp. 1261–1265, Dec. 1981.
- [6] E. Yamashita, K. Atsuki, and T. Ueda, "An approximate dispersion formula of microstrip lines for computer-aided design of microwave integrated circuits," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-27, pp. 1036–1038, Dec. 1979.

- [7] E. J. Denlinger, "A frequency dependent solution for microstrip transmission lines," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-19, pp. 30–39, Jan. 1971.
- [8] E. Hammerstad and O. Jensen, "Accurate modes for microstrip computer-aided design," in *1980 G-MTT Int. Microwave Symp. Dig.*, 1980, pp. 407–409.
- [9] M. Kobayashi, "Analysis of the microstrip and the electrooptic light modulator," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-26, pp. 119–126, Feb. 1978.
- [10] M. Kobayashi and R. Terakado, "New view on an anisotropic medium and its application to transformation from anisotropic to isotropic problems," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-27, pp. 769–775, Sept. 1979.
- [11] M. Kobayashi and R. Terakado, "Accurately approximate formula of effective filling fraction for microstrip line with isotropic substrate and its application to the case with anisotropic substrate," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-27, pp. 776–778, Sept. 1979.
- [12] M. Kobayashi, "Important role of inflection frequency in the curves of dispersion property of microstrip lines," pp. 2057–2059, this issue.

### Important Role of Inflection Frequency in the Dispersive Property of Microstrip Lines

MASANORI KOBAYASHI, MEMBER, IEEE

**Abstract**—It is represented that the inflection frequency  $f_i$  has the important role in the dispersive property of microstrip lines. This  $f_i$  is related to the coupling between the TEM mode and the  $TM_0$  mode. Using  $f_i$ , an approximate dispersion formula is derived by improving Schneider's formula. The results obtained by the present formula are compared with the other available results; good agreement is seen.

A microstrip transmission line is an essential part of an integrated circuit which is a fundamental component in modern microwave devices. With its increasing use at higher frequencies, a number of workers have theoretically studied the dispersive properties of microstrip lines [1]–[5] (good bibliographies are given in [6] and [10]). However, their analyses usually require a complicated computer program and, in some cases, enormous computing time. Recently, their results have been compared [6]. On the other hand, the computer-aided design of microstrip circuits requires accurate and reliable information on the dispersive behavior. A few approximate equations satisfying these requirements have been formulated [8]–[15], [17].

In a microstrip geometry, transverse TM- and TE-wave modes exist. Only even-order TM surface-wave modes and odd-order TE surface-wave modes are possible [18]. The  $TM_0$  mode is dominant since it has a zero frequency cutoff [18], while higher order

Manuscript received June 3, 1982; revised July 15, 1982.

The author is with the Department of Electrical Engineering, Faculty of Engineering, Ibaraki University, 4-12-1 Nakanarusawa-Machi, Hitachi, Ibaraki, Japan.

modes have cutoff frequencies beyond the range of interest, therefore, having a negligible effect [9]. Therefore, the important modal coupling is that between the TEM microstrip-line mode and the  $TM_0$  mode [7], [9], [16]. In this short paper, the inflection frequency  $f_i$  is related to this coupling. An approximate dispersion formula is derived by using this  $f_i$ .

We consider the microstrip line of strip width  $w$ , substrate thickness  $h$ , and permittivity of substrate  $\epsilon = \epsilon^* \epsilon_0$  ( $\epsilon^*$  is the relative dielectric constant,  $\epsilon_0$  is the permittivity of vacuum). Using the wave equations and the continuity conditions for fields at the interface of the different media in such a line [18], we can derive a coupling frequency  $f_{K, TM}$  defined as that frequency at which the phase velocity of the TEM mode equals the phase velocity of the  $TM_0$  mode, and a coupling frequency  $f_{K, TE}$  between the TEM mode and the  $TE_1$  mode

$$f_{K, TM} = \frac{v_0 \tan^{-1} \left( \epsilon^* \sqrt{\frac{\epsilon_{eff,0}^* - 1}{\epsilon^* - \epsilon_{eff,0}^*}} \right)}{2\pi h \sqrt{\epsilon^* - \epsilon_{eff,0}^*}} \quad (1)$$

$$f_{K, TE} = \frac{v_0 \left( \frac{\pi}{2} + \tan^{-1} \sqrt{\frac{\epsilon_{eff,0}^* - 1}{\epsilon^* - \epsilon_{eff,0}^*}} \right)}{2\pi h \sqrt{\epsilon^* - \epsilon_{eff,0}^*}} \quad (2)$$

where  $\epsilon_{eff,0}^*$  denotes the effective dielectric constant at zero frequency and is obtained also by

$$\epsilon_{eff,0}^* = 1 + q(\epsilon^* - 1) \quad (3)$$

where  $q$  is the effective filling fraction which has been tabulated for a wide range of  $\epsilon^*$  and  $w/h$  [20], and  $v_0$  denotes the speed of light in free space. Accurate approximate formulas have been obtained [19], [21]. We have the following limits for  $f_{K, TM}$ ,  $f_{K, TE}$  since  $q$  lies between 0.5 (when  $w/h \rightarrow 0$ ) and 1 (when  $w/h \rightarrow \infty$ ) [19], [20], [21]

$$1.5(q = 0.5, \epsilon^* \rightarrow \infty) \leq \frac{f_{K, TE}}{f_{K, TM}} \leq 3(q = 1). \quad (4)$$

We also note that the  $TM_0$  mode is dominant since  $f_{K, TM}$  is always smaller than  $f_{K, TE}$ .

Schneider [10] showed the four properties which have to be satisfied by the normalized phase velocity

$$\tilde{v}_p = \frac{v}{v_0} = \frac{1}{\sqrt{\epsilon_{eff,f}^*}}. \quad (5)$$

They are as follows: 1)  $\tilde{v}_p$  is a monotonically decreasing function of  $f$ ; 2)  $\tilde{v}_p = 1/\sqrt{\epsilon_{eff,0}^*}$ ,  $\partial \tilde{v}_p / \partial f = 0$  when  $f = 0$ ; 3)  $\tilde{v}_p \rightarrow 1/\sqrt{\epsilon^*}$ ,  $\partial \tilde{v}_p / \partial f \rightarrow 0$  when  $f \rightarrow \infty$ ; 4)  $\partial^2 \tilde{v}_p / \partial f^2 = 0$  in the vicinity of the cutoff frequency  $f_{C, TE}$  of the  $TE_1$  surface wave. They have been checked on the available data obtained theoretically or experimentally by many workers. The first three properties have been confirmed. However, the last property could not be confirmed in all cases. The curve of this dispersion formula which is obtained by replacing  $f_K$  with  $f_{C, TE}$  in (8) and (9) has an inflection frequency  $f_i$  as follows:

$$f_i = f_{C, TE} / \sqrt{3} \quad (6)$$

where

$$f_{C, TE} = v_0 / (4h\sqrt{\epsilon^* - 1}). \quad (7)$$

We notice in (7) that  $f_{C, TE}$  is a function of only  $h$  and  $\epsilon^*$ , but not of  $w$ . Therefore,  $f_i$  is also independent of  $w$ . On the other hand,

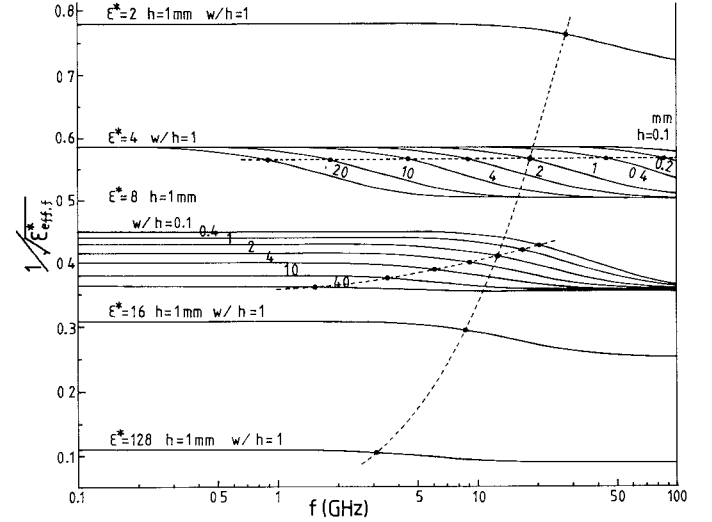


Fig. 1. Dispersion behavior of the approximate formula (8) for some variations of parameters  $\epsilon^*$ ,  $h$ , and  $w/h$ .

we have observed from available data that  $f_i$  must be a function of not only  $h$  and  $\epsilon^*$ , but also  $w$ . However, the author was interested in Schneider's formula with the simple form and considered improving it.

The order of magnitude of  $\epsilon_{eff,f}^*$  is determined by  $\epsilon_{eff,0}^*$ , that is,  $w/h$  and  $\epsilon^*$  (see Fig. 1). On the other hand,  $f_i$  strongly influences the dispersive nature of the line (see Fig. 1,  $\bullet$  denotes  $f_i$ ). Therefore, the value of  $f_i$  is very important in deriving the approximate formula. The author believes that  $f_i$  ought to have a strong relation to  $f_{K, TM}$  shown in (1) since the  $TM_0$  mode is dominant. The author proposes the following approximate formula based upon the theoretical considerations given above:

$$\frac{1}{\sqrt{\epsilon_{eff,f}^*}} = \frac{\frac{1}{\sqrt{\epsilon^*}} \left( \frac{f}{f_K} \right)^2 + \frac{1}{\sqrt{\epsilon_{eff,0}^*}}}{\left( \frac{f}{f_K} \right)^2 + 1} \quad (8)$$

where

$$f_K = f_{K, TM} / \left( 1 + \frac{w}{h} \right) \quad (9)$$

and the denominator in (9) is only one corrective factor. Schneider's formula [10] is identical to the one obtained by replacing  $f_K$  with  $f_{C, TE}$  in (8).

The dispersion curve in (8) has the following inflection frequency  $f_i$  at which the second-order derivative of  $1/\sqrt{\epsilon_{eff,f}^*}$  with respect of  $f$  is zero:

$$f_i = f_K / \sqrt{3}. \quad (10)$$

The  $1/\sqrt{\epsilon_{eff,f}^*}$  at  $f_i$  has, for any  $\epsilon^*$ ,  $h$ , and  $w$ , the following value:

$$\frac{1}{\sqrt{\epsilon_{eff,f_i}^*}} = \frac{1}{4} \left( \frac{3}{\sqrt{\epsilon_{eff,0}^*}} + \frac{1}{\sqrt{\epsilon^*}} \right). \quad (11)$$

Fig. 1 shows the dispersive behavior of the present formula (8) for several variations of the parameters ( $\epsilon^*$ ,  $h$ , and  $w/h$ ). The dispersion curves are monotonically decreasing  $1/\sqrt{\epsilon_{eff,0}^*}$  (at  $f = 0$ ) to  $1/\sqrt{\epsilon^*}$  (at  $f = \infty$ ), due to an increase in frequency. On these curves,  $f_i$  is the point fallen down just a quarter of all variation

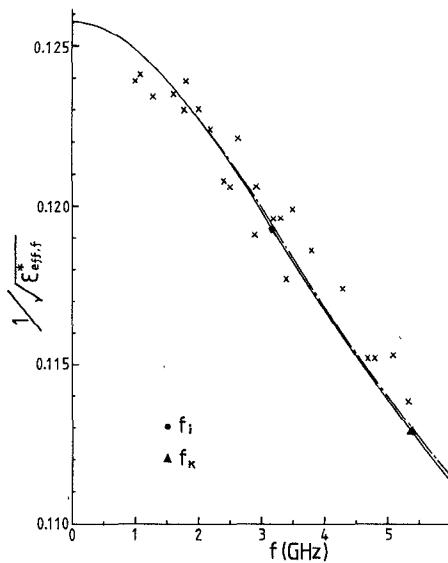


Fig. 2. Dispersion behavior of  $1/\sqrt{\epsilon_{\text{eff},f}^*}$ .  $\epsilon^* = 104$ ,  $h = 1.27$  mm,  $w/h = 0.6$ ,  $q = 0.604191$ ,  $f_i = 3.3186$ ,  $f_K = 5.7480$ ,  $f_{K,TM} = 9.1968$ ,  $f_{C,TE} = 5.8148$  GHz. — the present formula, ——— Schneider's formula [10].  $\times$  Hartwig *et al.* [16].

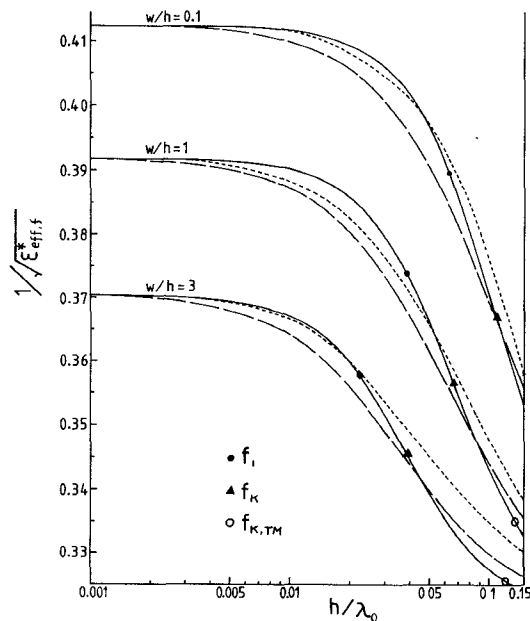


Fig. 3. Dispersion behavior of  $1/\sqrt{\epsilon_{\text{eff},f}^*}$ .  $\epsilon^* = 9.7$ ;  $q = 0.560246$ ,  $f_i = 18.945$ ,  $f_K = 32.814$ ,  $f_{K,TM} = 36.095$ ,  $f_{C,TE} = 25.409$  GHz for  $w/h = 0.1$ ;  $q = 0.633505$ ,  $f_i = 11.512$ ,  $f_K = 19.940$ ,  $f_{K,TM} = 39.881$ ,  $f_{C,TE} = 25.409$  GHz for  $w/h = 1$ ;  $q = 0.722898$ ,  $f_i = 6.6845$ ,  $f_K = 11.577$ ,  $f_{K,TM} = 46.311$ ,  $f_{C,TE} = 25.409$  GHz for  $w/h = 3$   $h/\lambda_0 = f(\text{GHz})/299.7925$ . — the present formula (8), ——— Yamashita's formula [14], ——— Kowalski *et al.* [1].

from the  $1/\sqrt{\epsilon_{\text{eff},0}^*}$ . Therefore, we can approximately draw the curves of  $1/\sqrt{\epsilon_{\text{eff},f}^*}$  versus  $f$  if  $1/\sqrt{\epsilon_{\text{eff},0}^*}$ ,  $1/\sqrt{\epsilon^*}$ , and  $f_i$  are given. We can graphically confirm in Fig. 1 the important results; the larger is  $\epsilon^*$ ,  $w/h$ , or  $h$ , the lower is the inflection frequency  $f_i$ . Thus, we can understand the important roles of  $f_i$ .

Figs. 2 and 3 compare the results of the present formula with the other available results [1], [10], [14], [16]. The values of  $1/\sqrt{\epsilon_{\text{eff},0}^*}$  at the zero frequency were calculated by the Green's function technique [20] which has a very high degree of accuracy.

The good agreement of this formula (solid line) and Schneider's formula (dot-dash-line) [10] as shown in Fig. 2 is due to the good agreement of  $f_K = 5.748(\text{GHz})$  and  $f_{C,TE} = 5.814(\text{GHz})$  that is, both the inflection frequencies  $f_i$  are nearly equal. In Fig. 3, the dotted lines denotes the theoretical results by Kowalski *et al.* [1], the broken lines denotes the results of Yamashita's formula [4], and the solid lines the results of the present formula. Good agreement can be seen. We find that the present formula has good agreement with the theoretical results near  $f_i$ .

In summary, the important roles of  $f_i$  are represented by letting  $f_i$  be related to the coupling between the  $\text{TEM}_i$  mode and the  $\text{TM}_0$  mode  $f_{K,TM}$ . This result is applied to the article [22].

#### ACKNOWLEDGMENT

The author thanks Prof. Naito, Tokyo Institute of Technology, for providing copies of references [16], [17].

#### REFERENCES

- [1] G. Kowalski and R. Pregla, "Dispersion characteristics of single and coupled microstrips," *AEU*, vol. 26, pp. 276–280, June 1972.
- [2] T. Itoh and R. Mittra, "Spectral-domain approach for calculating the dispersion characteristics of microstrip lines," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-21, pp. 496–499, July 1973.
- [3] E. Yamashita and K. Atsuki, "Analysis of microstrip-like transmission lines by non-uniform discretization of integral equations," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-24, pp. 195–200, Apr. 1976.
- [4] Y. Fujiki, Y. Hayashi, and M. Suzuki, "Analysis of strip transmission lines by iteration method," *Trans. Inst. Electron. Comm. Eng.*, (Japan), vol. 55-B, pp. 212–219, May 1972, (in Japanese).
- [5] Y. Hayashi and T. Kitazawa, "Analysis of microstrip transmission line on a sapphire substrate," *Trans. Inst. Electron. Comm. Eng.*, (Japan), vol. 62-B, pp. 596–602, June 1979, (in Japanese).
- [6] E. F. Kuester and D. C. Chang, "An appraisal of methods for computation of the dispersion characteristics of open microstrip," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-27, pp. 691–694, July 1979.
- [7] G. D. Vendelin, "Limitations on stripline Q," *Microwave J.*, vol. 13, pp. 63–69, May 1970.
- [8] E. J. Denlinger, "A frequency dependent solution for microstrip transmission lines," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-19, pp. 30–39, Jan. 1971.
- [9] O. P. Jain, V. Makios, and W. J. Chudobiak, "Coupled-mode model of dispersion in microstrip," *Electron. Lett.*, vol. 7, pp. 405–407, July 1971.
- [10] M. V. Schneider, "Microstrip dispersion," *Proc. IEEE*, vol. 60, pp. 144–146, Jan. 1972.
- [11] W. J. Getsinger, "Microstrip dispersion model," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-21, pp. 34–39, Jan. 1973.
- [12] H. J. Carlin, "A simplified circuit model for microstrip," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-21, pp. 589–591, Sept. 1973.
- [13] T. C. Edwards and R. P. Owens, "2–18-GHz dispersion measurements on 10–100- $\Omega$  microstrip lines on sapphire," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-24, pp. 506–513, Aug. 1976.
- [14] E. Yamashita, K. Atsuki, and T. Ueda, "An approximate dispersion formula of microstrip lines for computer-aided design of microwave integrated circuits," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-27, pp. 1036–1038, Dec 1979.
- [15] E. Yamashita, K. Atsuki, and T. Hirahata, "Microstrip dispersion in a wide-frequency range," *IEEE Microwave Theory Tech.*, vol. MTT-29, pp. 610–611, June 1981.
- [16] C. P. Hartwig, D. Massé, and R. A. Pucel, "Frequency dependent behavior of microstrip," in *1968 G-MTT Int. Microwave Symp. Dig.*, 1968, pp. 110–116.
- [17] E. Hammerstad and O. Jensen, "Accurate models for microstrip computer-aided design," in *1980 G-MTT Int. Microwave Symp. Dig.*, pp. 407–409, 1980.
- [18] R. E. Collin, *Field Theory of Guided Waves*. New York: (McGraw-Hill, 1960, p. 470.
- [19] H. A. Wheeler, "Transmission-line properties of a strip on a dielectric sheet on a plate," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-25, pp. 631–647, Aug 1977.
- [20] M. Kobayashi, "Analysis of the microstrip and the electrooptic light modulator," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-26, pp. 119–126, Feb 1978.
- [21] M. Kobayashi and R. Terakado, "Accurately approximate formula of effective filling fraction for microstrip line with isotropic substrate and its application to the case with anisotropic substrate," *IEEE Trans. Microwave Theory Tech.*, vol. MTT-27, pp. 776–778, Sept. 1979.
- [22] M. Kobayashi, "Frequency dependent characteristics of microstrip on anisotropic substrates," pp. 2054–2057, this issue.